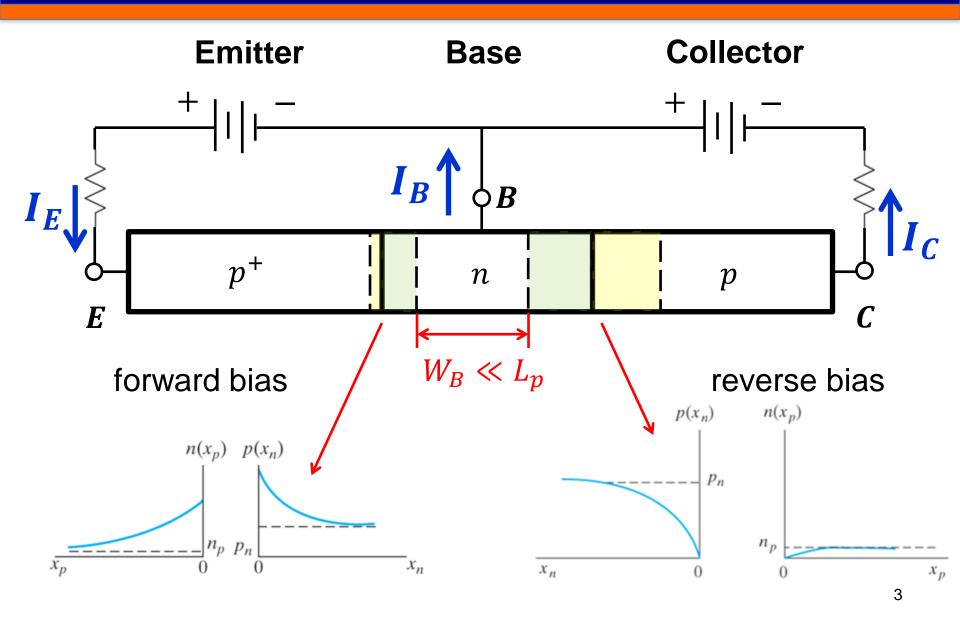
# ECE 340 Lecture 39 Semiconductor Electronics

Spring 2022 10:00-10:50am Professor Umberto Ravaioli Department of Electrical and Computer Engineering 2062 ECE Building

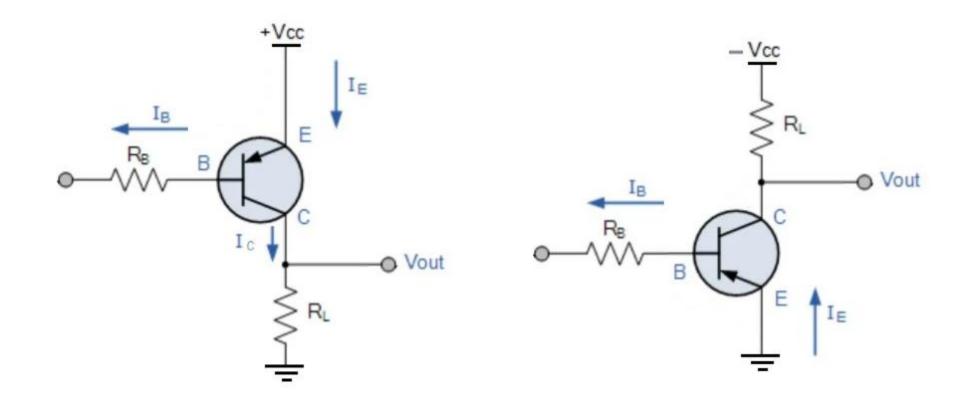
### Today's Discussion

 Introduction to the Bipolar Junction Transistor (BJT)

### Normal mode operation-common emitter

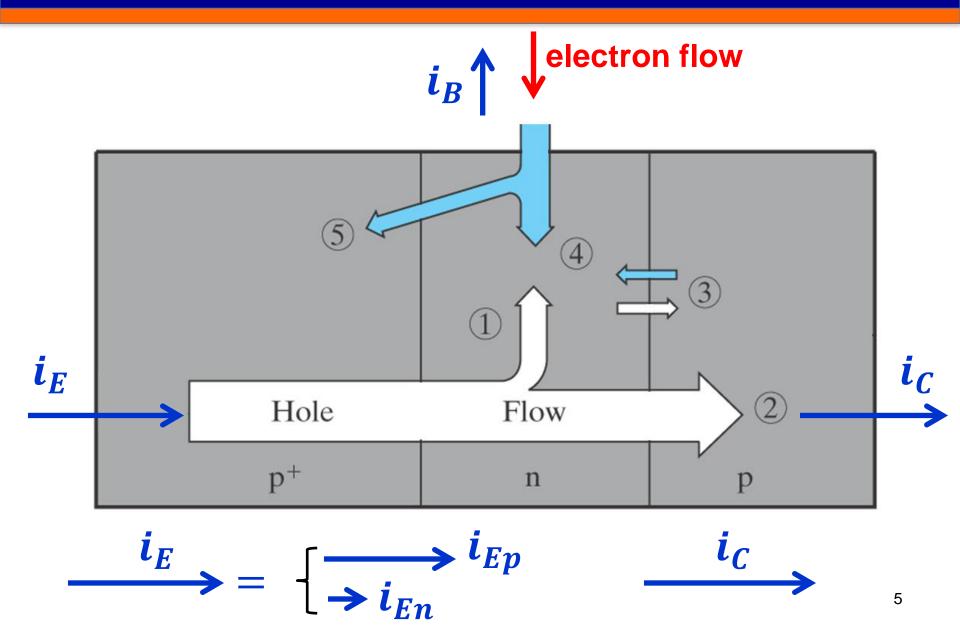


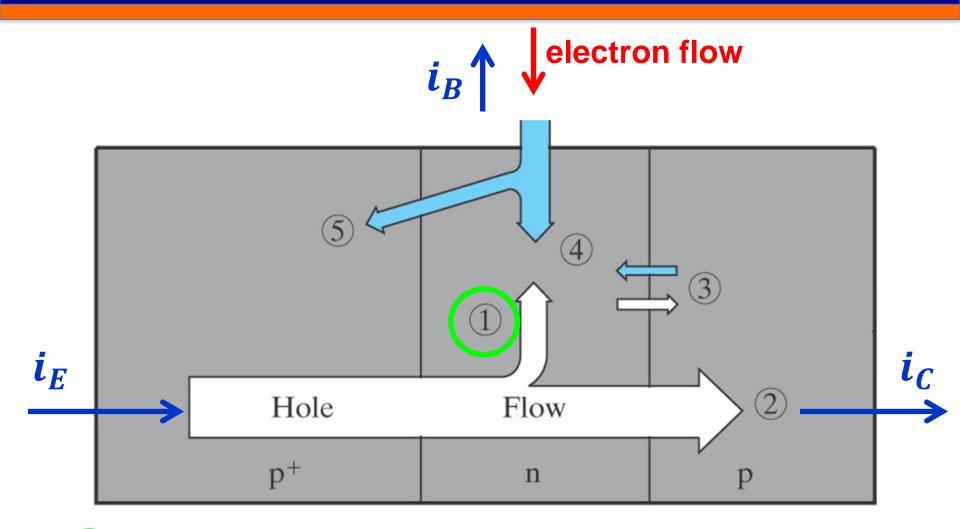
### Normal mode operation-common emitter



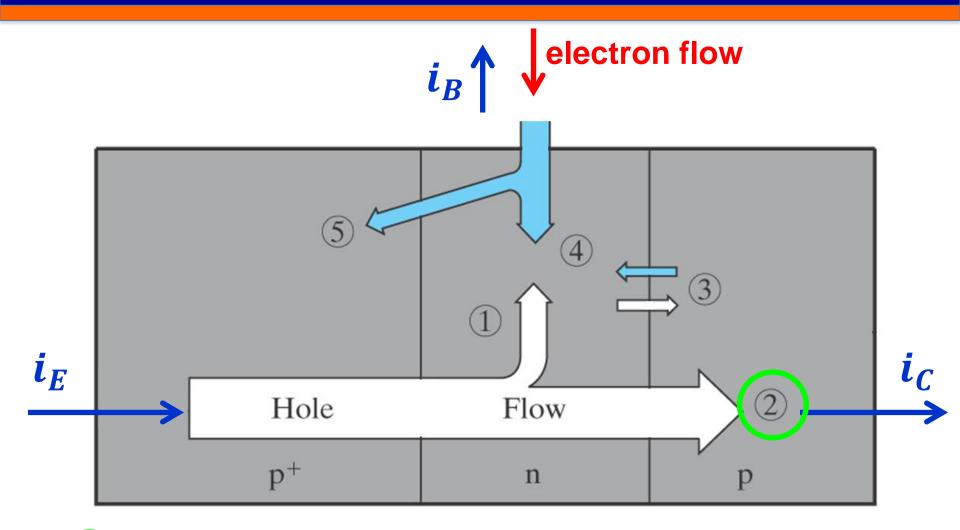
circuit schematics for p-n-p transistors

#### Let's review again carrier flow in BJT

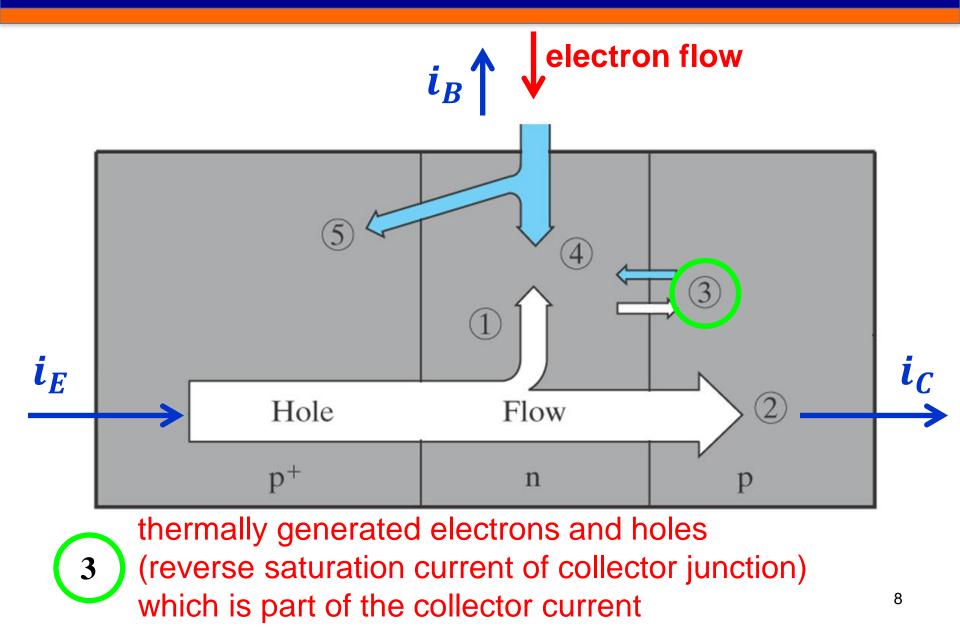


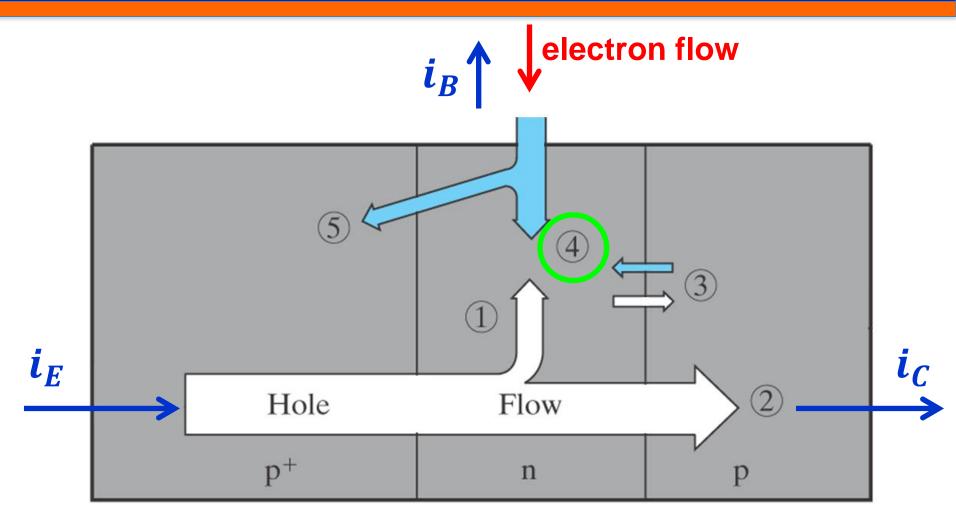


1 ) injected holes lost to recombination in the base



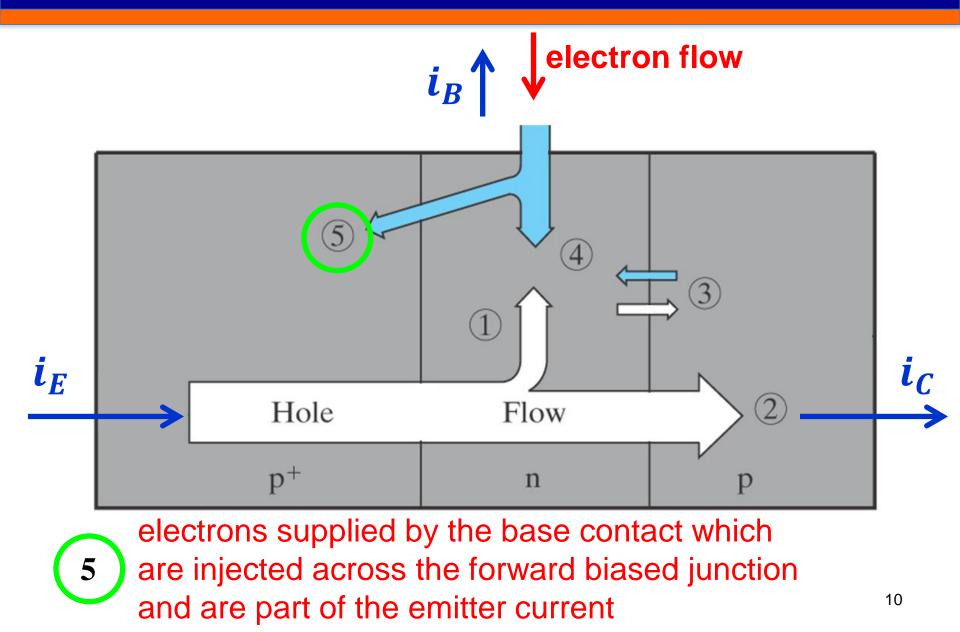
injected holes reaching reverse-biased collector junction



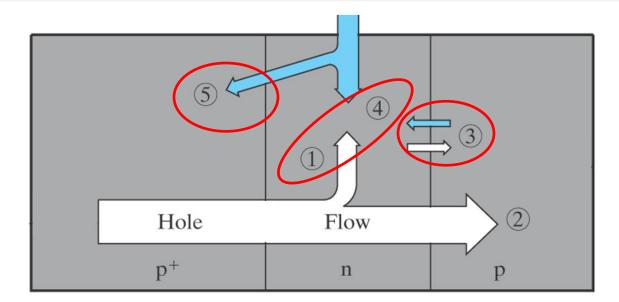


4

electrons supplied by the base contact which recombine with holes in the base neutral region

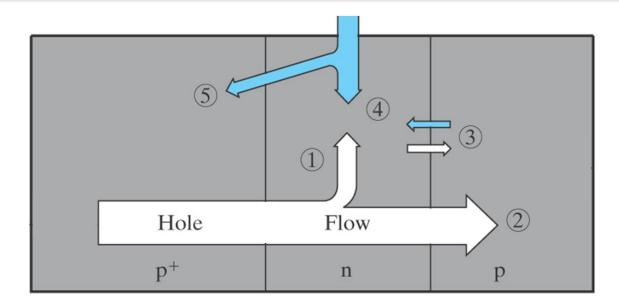


### Base current physical mechanisms



- recombination of injected holes even if  $W_B \ll L_p$
- injection of electrons into the emitter
- thermally generated electrons in the reverse biased junction are swept into the base reducing the supply from the base contact

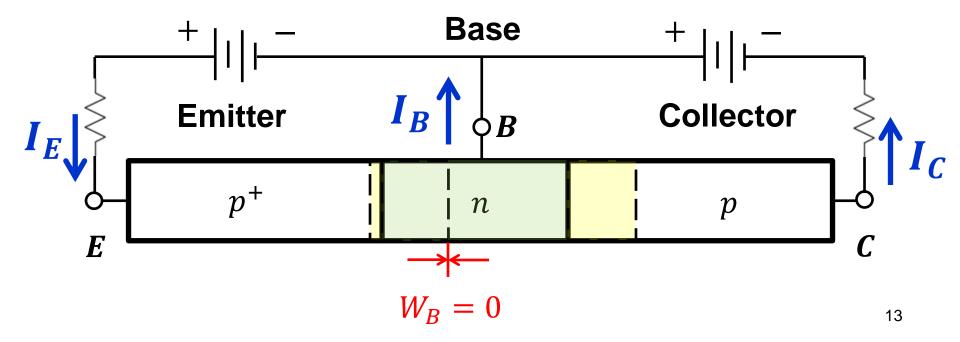
#### For a well-designed BJT



- Holes injected by the emitter into the base are collected as much as possible ( $I_E - I_C$  very small)  $\rightarrow$  we need base with narrow width and long hole lifetime so that  $W_B \ll L_p = \sqrt{D_p \tau_p}$
- Current crossing the emitter should consist almost entirely of holes → we need high doping in emitter with respect to base doping (e.g., p<sup>+</sup>-n emitter-base junction)

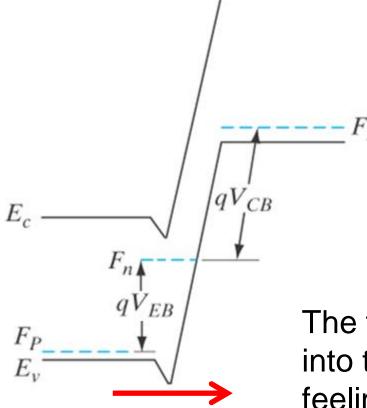
## Base limitations: punch-through

- If the base region is too thin and if the base doping is too light, at the desired voltages the depletions from the two junctions may meet, resulting in "punch-through".
- In such conditions, the base current is no longer able to control the emitter current.



## Base limitations: punch-through

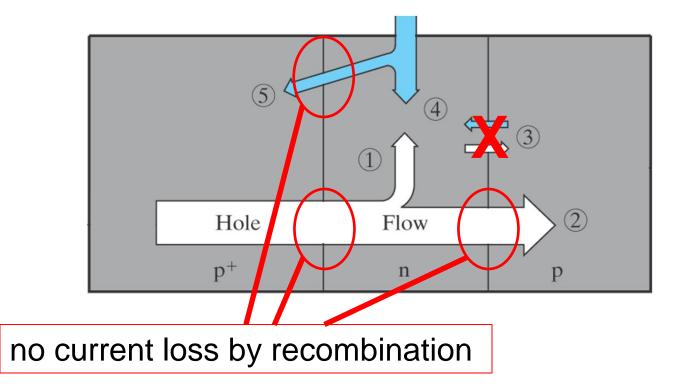
• The band diagram might look approximately like this:



The flow of holes goes directly into the collector region without feeling the influence of the base

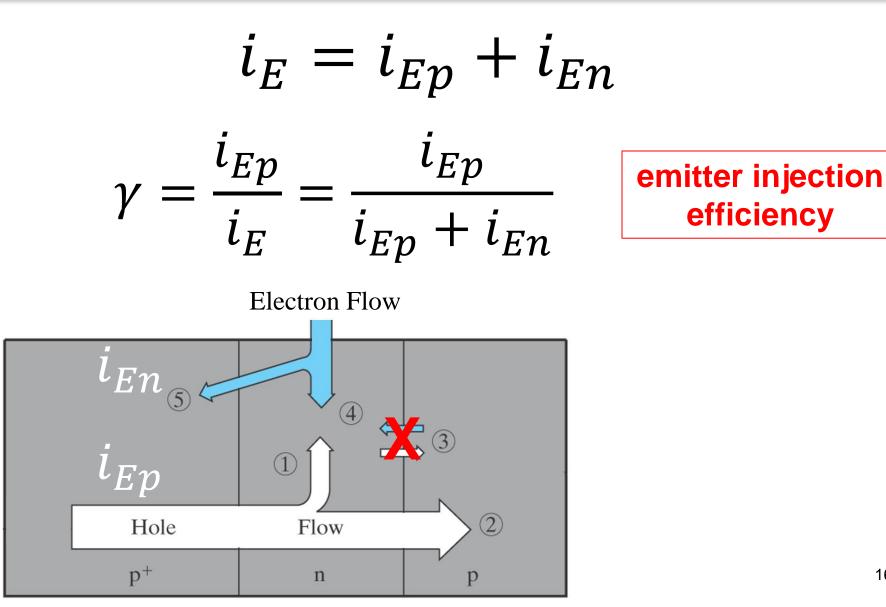
## **Amplification with BJT**

• The control input is the base current  $i_B$ 

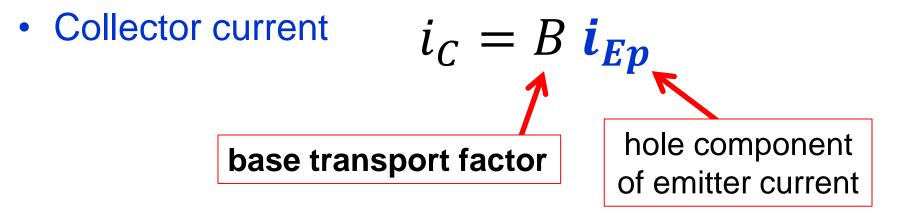


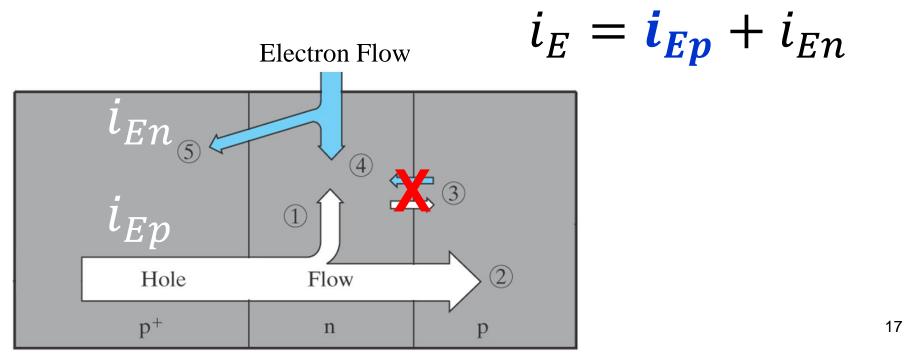
- Simplifying Assumptions:
  - Neglect reverse saturation current at collector
  - Neglect recombination in the depletion regions<sup>15</sup>

#### Amplification with BJT – injection efficiency



#### Amplification with BJT – base transport factor





### For an efficient transistor – 1

 Emitter injection efficiency → unity emitter current is mostly holes

$$\gamma = \frac{i_{Ep}}{i_E} = \frac{i_{Ep}}{i_{Ep} + i_{En}} = \frac{1}{1 + \frac{i_{En}}{i_{Ep}}}$$
$$\gamma \to 1$$
$$\frac{i_{En}}{i_{Ep}} \to 0$$
$$We need$$
$$i_{En} \ll i_{Ep}$$

### For an efficient transistor – 2

 Base transport factor → unity most of the holes make it to the collector

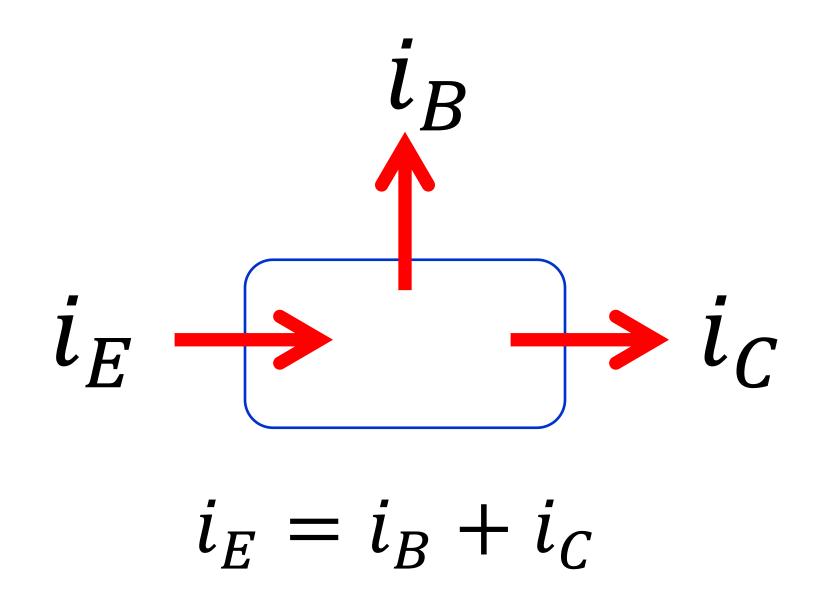
Minimize recombination in the base so that

$$i_C \approx i_{Ep}$$

#### Putting it all together

$$\frac{i_{C}}{i_{E}} = \frac{Bi_{Ep}}{i_{Ep} + i_{En}} = B\gamma = \alpha \frac{\text{current transfer}}{\text{ratio}}$$

#### **Base-to-Collector Amplification**



### **Base-to-Collector Amplification**



$$i_B = i_E - i_C = i_{En} + i_{Ep} - Bi_{Ep}$$
  
 $i_B = i_{En} + i_{Ep}(1 - B)$ 

B = fraction of holes that make it across the base (1 - B) is the fraction of holes that recombine

#### **Base-to-Collector Amplification**

$$\frac{i_{C}}{i_{B}} = \frac{B \ i_{Ep}}{i_{En} + i_{Ep}(1 - B)} = \qquad \begin{array}{c} \text{multyply by} \\ \frac{(i_{En} + i_{Ep})}{(i_{En} + i_{Ep})} = 1 \\ \\ B \ \frac{i_{Ep}}{i_{En} + i_{Ep}} \leftarrow = \gamma \\ \hline \frac{i_{En}}{i_{En} + i_{Ep}} + \frac{i_{Ep}}{i_{En} + i_{Ep}} (1 - B) \\ \hline B \gamma \end{array} = 1 \\ \\ \hline \frac{B \gamma}{i_{En} + i_{Ep}} \leftarrow B \left[ \frac{i_{Ep}}{i_{En} + i_{Ep}} \right] \leftarrow B \left[ \frac{i_{Ep}}{i_{En} + i_{Ep}} \right] \leftarrow \left[ -\frac{\beta}{i_{En} + i_{Ep}} \right] \leftarrow \left[ -\frac{\beta}{i_{En} + i_{Ep}} \right]$$

### Base-to-Collector Amplification Factor $\beta$

$$i_{C} = \alpha \qquad \begin{array}{c} current \ transfer \\ ratio \end{array}$$

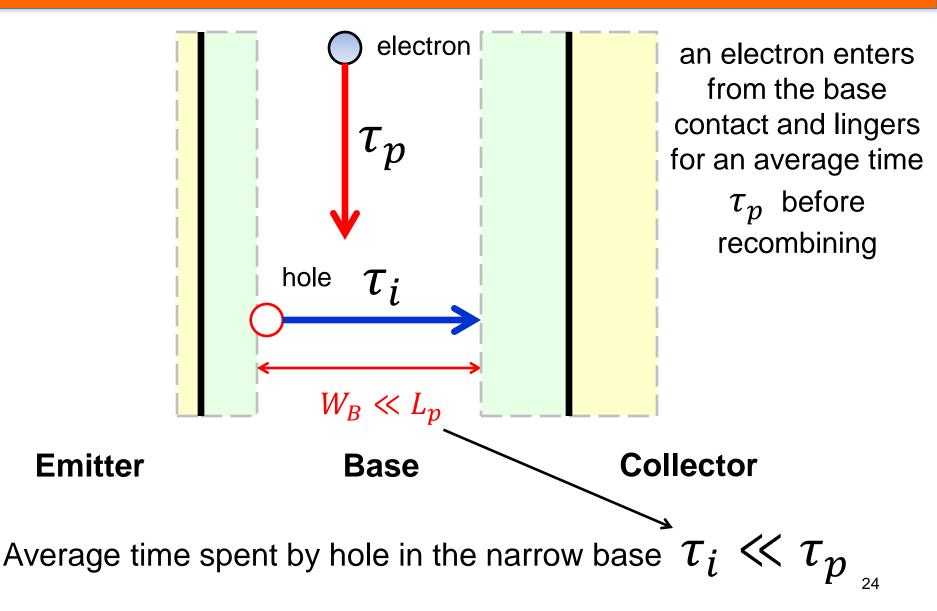
$$\frac{i_{C}}{i_{B}} = \frac{B\gamma}{1 - B\gamma} = \frac{\alpha}{1 - \alpha} = \beta$$

Since  $\alpha$  is close unity,  $\beta$  can be quite large

- $\alpha = 0.9 \rightarrow \beta = 9$
- $\alpha = 0.95 \rightarrow \beta = 19$
- $\alpha = 0.99 \rightarrow \beta = 99$

 $\alpha = 0.999 \rightarrow \beta = 999$ 

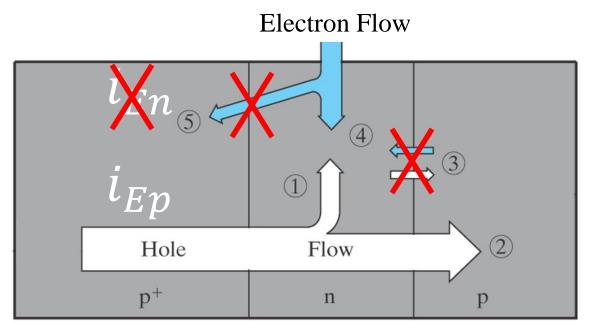
#### Hole transit time in the base



#### Charge storage in the base

At steady-state there are excess electrons and holes in the base and for charge neutrality  $Q_n = Q_p$ 

## Assume $\tau_n = \tau_p$ , perfect emitter injection efficiency ( $\gamma = 1$ ) and negligible saturation current



### Charge storage in the base

$$i_C = \frac{Q_p}{\tau_i}$$

$$i_B = \frac{Q_n}{\tau_n} = \frac{Q_n}{\tau_p}$$

$$Q_p = i_C \tau_i \qquad Q_n = i_B \tau_p \qquad Q_n = Q_p$$

$$i_C \tau_i = i_B \tau_n$$

For each electron entering from the base contact, a number  $\tau_p/\tau_i$  of holes goes from emitter to collector maintaining charge neutrality

$$\frac{i_C}{i_B} = \frac{\tau_p}{\tau_i} = \beta$$

## Mathematical analysis of the *p-n-p* BJT

- Some simplifying assumptions are necessary in order to develop a manageable model which is general and valid for general bias conditions:
- 1. Negligible drift in the base region (holes move by diffusion)
- 2. Emitter injection efficiency  $\gamma = 1$  (emitter is highly doped *p*+)
- 3. Reverse saturation current at the collector is negligible
- 4. Uniform cross-sectional area A (1-D model)
- 5. Steady-state conditions

We are going to focus on the significance of the results and on physical understanding of physical behavior.

Details of the analytical solution for the 1-D model BJT are outlined in the posted handout and are left as optional reading for the interested students.

Actual complete simulations of realistic devices are carried out by numerical solution of the coupled system of semiconductor equations consisting of:

- continuity equations for electrons and holes based on the drift-diffusion current model
- Poisson equation to obtained self-consistent space dependent electric fields

#### Results obtained from analytical solution

$$I_{Ep} = qA \frac{D_p}{L_p} \left[ \Delta p_E \operatorname{ctnh} \frac{W_B}{L_P} - \Delta p_C \operatorname{csch} \frac{W_B}{L_P} \right]$$

$$I_{C} = qA \frac{D_{p}}{L_{p}} \left[ \Delta p_{E} \operatorname{csch} \frac{W_{B}}{L_{P}} - \Delta p_{C} \operatorname{ctnh} \frac{W_{B}}{L_{P}} \right]$$

$$I_B = qA \frac{D_p}{L_p} \left[ (\Delta p_E + \Delta p_C) \tanh \frac{W_B}{2L_P} \right]$$

#### For the narrow base diode

$$I_p(x_n = 0) = qA \frac{D_p}{L_p} \Delta p_n \operatorname{ctnh} \frac{\ell}{L_p}$$
$$I_p(x_n = \ell) = qA \frac{D_p}{L_p} \Delta p_n \operatorname{csch} \frac{\ell}{L_p}$$
$$I_n(\operatorname{recomb}) = qA \frac{D_p}{L_p} \Delta p_n \operatorname{tanh} \frac{\ell}{2L_p}$$

With  $\Delta p_{c} pprox 0$  essentially the same result obtained for BJT

### For the curious ones:

 Video by Bill Hammack on the first transistor invented by Bardeen and Brattain at Bell Labs (point-contact transistor)

https://www.youtube.com/watch?v=RdYHljZi7ys

The book by Shockley contains an extensive description of the point-contact transistor, based on metalsemiconductor junctions rather than p-n junctions <u>https://archive.org/details/ElectronsAndHolesInSemiconductors</u>

 AT&T Archives video: Genesis of the Transistor: <u>https://www.youtube.com/watch?v=WiQvGRjrLnU</u>